

# Midterm Review

Midterm only covers material from lectures and HWs

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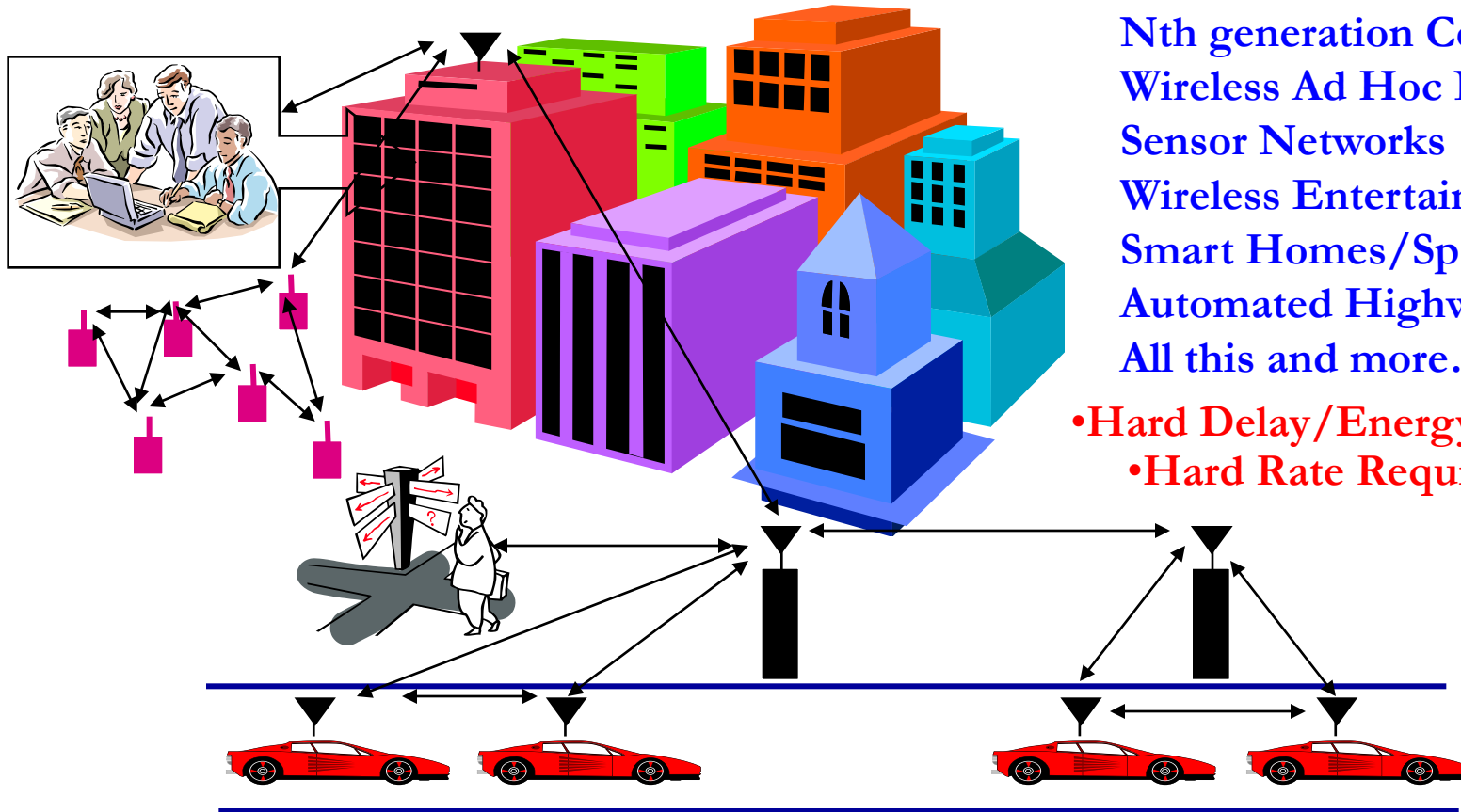
- **Overview of Wireless Systems:**
  - Nothing from Chapter 1 is on the MT
- **Signal Propagation and Channel Models**
  - Chapter 2.1-2.4, 2.6-2.10
- **Modulation and Performance Metrics**
  - Chapter 3.1,3.2.1-3.2.2, 3.3
- **Fundamental Capacity Limits**
  - Chapter 4
- **Impact of Channel on Performance**
  - Chapter 6
- **Diversity Techniques**
  - Chapter 7.1,7.2.1-7.2.2,7.2.4,7.3.1,7.4.1

# Future Wireless Networks

## *Ubiquitous Communication Among People and Devices*

Wireless Internet access  
Nth generation Cellular  
Wireless Ad Hoc Networks  
Sensor Networks  
Wireless Entertainment  
Smart Homes/Spaces  
Automated Highways  
All this and more...

- Hard Delay/Energy Constraints
- Hard Rate Requirements



# Design Challenges

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- **Wireless channels are a difficult and capacity-limited broadcast communications medium**
- **Traffic patterns, user locations, and network conditions are constantly changing**
- **Applications are heterogeneous with hard constraints that must be met by the network**
- **Energy, delay, and rate constraints change design principles across all layers of the protocol stack**

# Current/Futre Wireless Systems

- Current Systems
  - 4G Cellular Systems
  - 802.11b/a/n/ac Wireless LANs
  - Satellite Systems
  - Paging Systems
  - Bluetooth
  - Zigbee radios
- Emerging Systems (Can cover in bonus lecture)
  - Ad hoc/mesh wireless networks
  - Cognitive radio networks
  - Wireless sensor networks
  - Energy-harvesting radios
  - Distributed control networks
  - Communications/SP in Health, Bio-medicine, and Neuroscience

# Signal Propagation

- Path Loss

- Free space, 2-path,...
- Simplified model

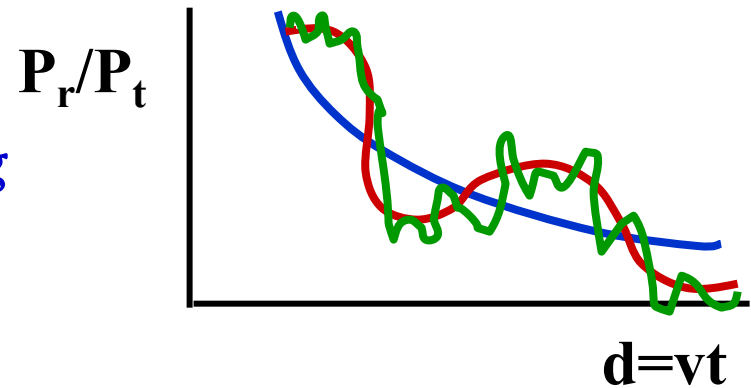
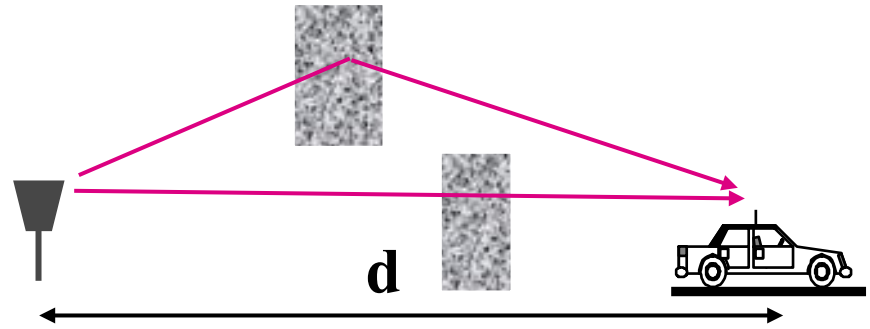
$$P_r = P_t K \left[ \frac{d_0}{d} \right]^\gamma, \quad 2 \leq \gamma \leq 8$$

- Shadowing

- dB value is Gaussian
- Find path loss exponent and shadow STD by curve fitting

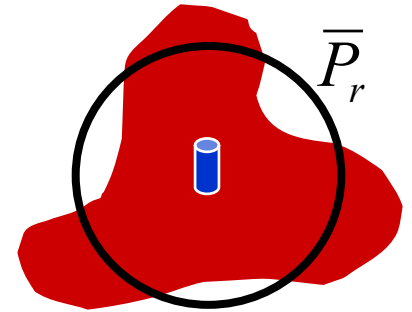
- Multipath

- Ray tracing
- Statistical model

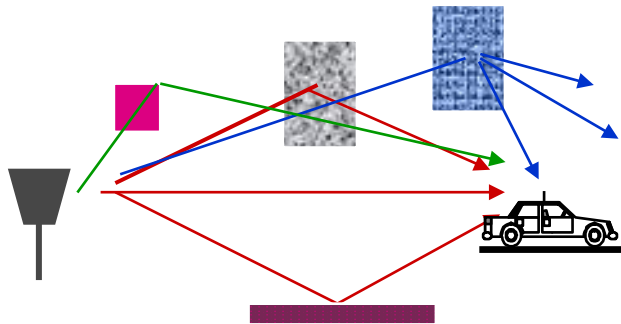


# Outage Probability and Cell Coverage Area

- Path loss: circular cells
- Path loss+shadowing: amoeba cells
  - Tradeoff between coverage and interference
- Outage probability
  - Probability received power below given minimum
- Cell coverage area
  - % of cell locations at desired power
  - Increases as shadowing variance decreases
  - Large % indicates interference to other cells



# Statistical Multipath Model



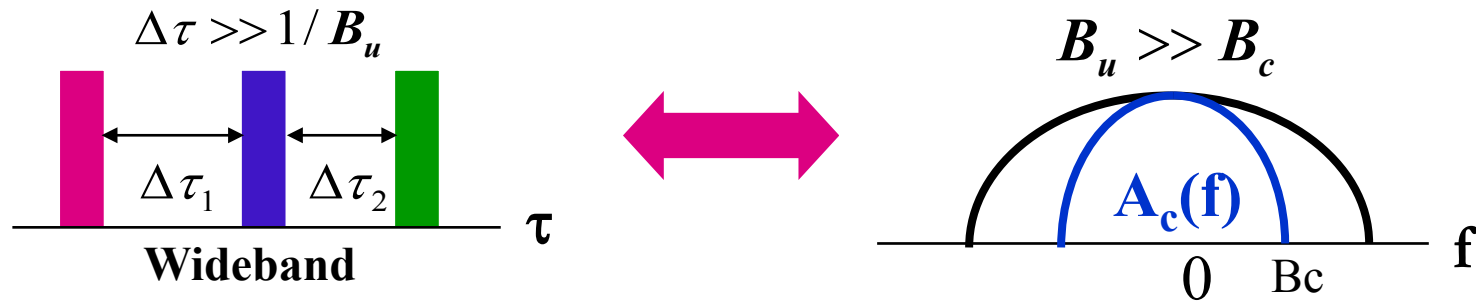
- Random # of multipath components, each with varying amplitude, phase, doppler, and delay
- Leads to time-varying channel impulse response

$$c(\tau, t) = \sum_{n=1}^N \alpha_n(t) e^{-j\phi_n(t)} \delta(\tau - \tau_n(t))$$

- Narrowband channel
  - No signal distortion, just a complex amplitude gain
  - Signal amplitude varies randomly (Rayleigh, Ricean, Nakagami).
  - 2<sup>nd</sup> order statistics (Bessel function), Average fade duration

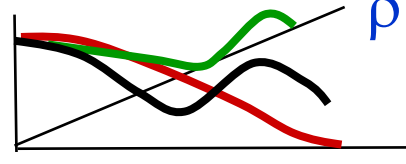
# Wideband Channels

- Individual multipath components resolvable
- True when time difference between components exceeds signal bandwidth



- Scattering function

$$s(\tau, \rho) = \mathcal{F}_{\Delta t}[A_c(\tau, \Delta t)]$$



Doppler Power Spectrum

Delay Power Spectrum

- Yields delay spread/coherence BW ( $\sigma_\tau \sim 1/B_c$ )
- Yields Doppler spread/coherence time ( $B_d \sim 1/T_c$ )



# Capacity of Flat Fading Channels

- Channel Capacity

- Maximum data rate that can be transmitted over a channel with arbitrarily small error

- Capacity of AWGN Channel:  $B \log_2[1+\gamma]$  bps

- $\gamma = P_r / (N_0 B)$  is the receiver SNR

- Capacity of Flat-Fading Channels

- Nothing known: capacity typically zero
- Fading Statistics Known (few results)
- Fading Known at RX (average capacity)

$$C = \int_0^{\infty} B \log_2(1 + \gamma) p(\gamma) d\gamma \leq B \log_2(1 + \bar{\gamma})$$

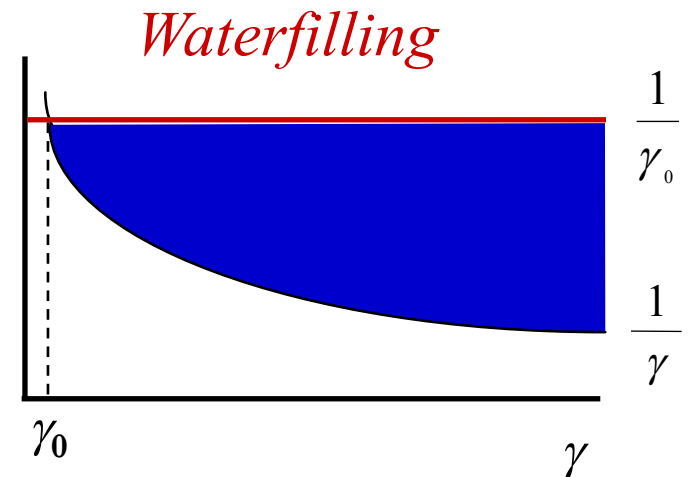
- Capacity in Flat-Fading:  $\gamma$  known at TX/RX

$$C = \max_{P(\gamma) : E[P(\gamma)] = \bar{P}} \int_0^{\infty} B \log_2 \left( 1 + \frac{\gamma P(\gamma)}{\bar{P}} \right) p(\gamma) d\gamma$$

- Optimal Rate and Power Adaptation

$$\frac{P(\gamma)}{\bar{P}} = \begin{cases} \frac{1}{\gamma_0} - \frac{1}{\gamma} & \gamma \geq \gamma_0 \\ \mathbf{0} & \text{else} \end{cases}$$

$$\frac{C}{B} = \int_{\gamma_0}^{\infty} \log_2 \left( \frac{\gamma}{\gamma_0} \right) p(\gamma) d\gamma.$$



- The instantaneous power/rate only depend on  $p(\gamma)$  through  $\gamma_0$

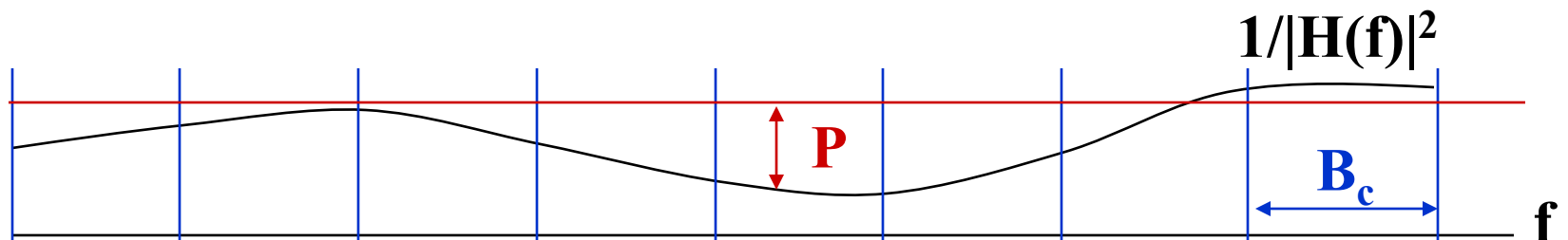
# Channel Inversion

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- Fading inverted to maintain constant SNR
- Simplifies design (fixed rate)
- Greatly reduces capacity
  - Capacity is zero in Rayleigh fading
- Truncated inversion
  - Invert channel above cutoff fade depth
  - Constant SNR (fixed rate) above cutoff
  - Cutoff greatly increases capacity
    - Close to optimal

# Frequency Selective Fading Channels

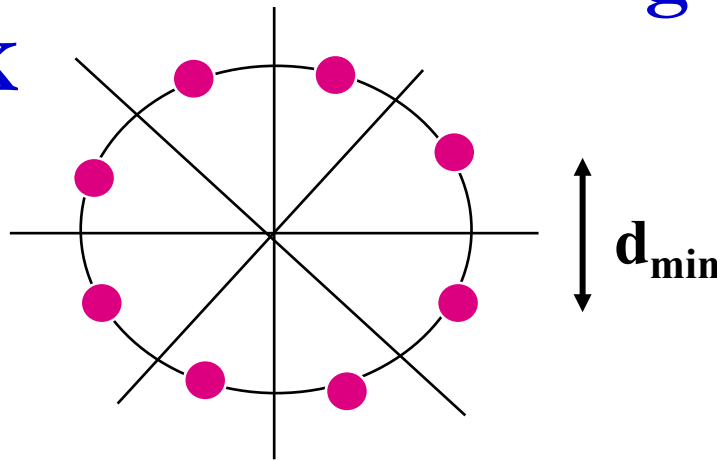
- For time-invariant channels, capacity achieved by water-filling in frequency
- Capacity of time-varying channel unknown
- Approximate by dividing into subbands
  - Each subband has width  $B_c$
  - Independent fading in each subband
  - Capacity is the sum of subband capacities



# Linear Modulation in AWGN: MPSK and MQAM

- ML detection induces decision regions

- Example: 8PSK



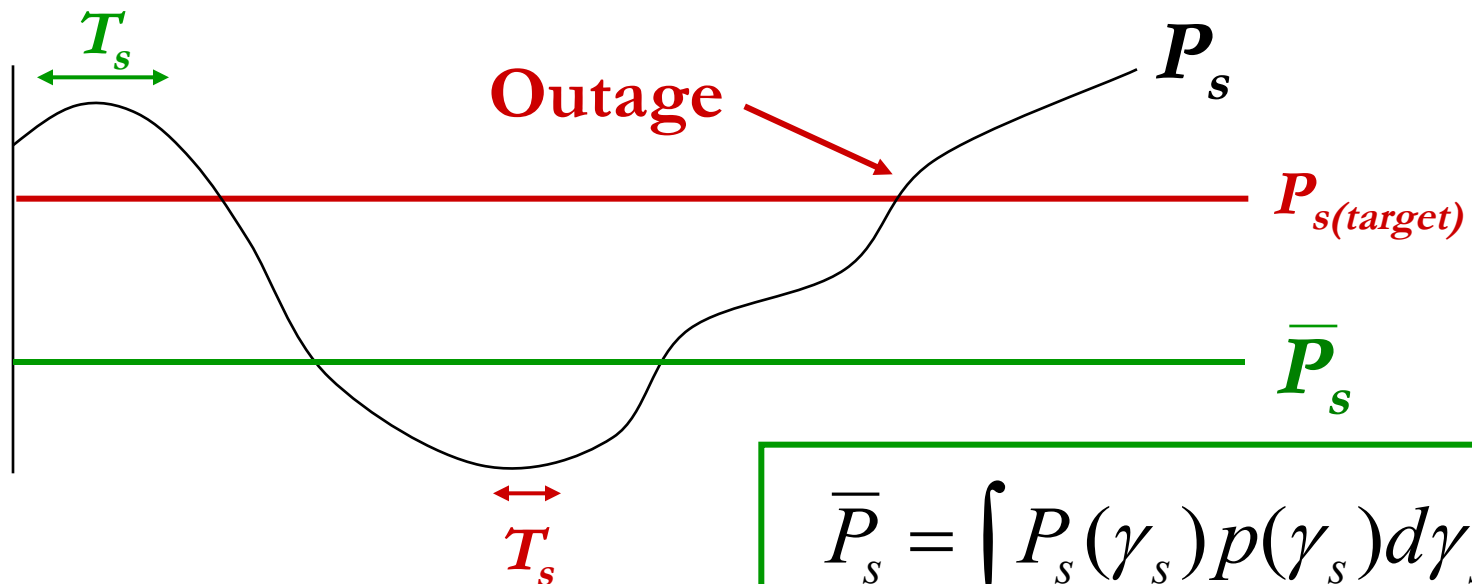
- $P_s$  depends on

- # of nearest neighbors
- Minimum distance  $d_{\min}$  (depends on  $\gamma_s$ )
- Approximate expression

$$P_s \approx \alpha_M Q\left(\sqrt{\beta_M \gamma_s}\right)$$

# Linear Modulation in Fading

- In fading  $\gamma_s$  and therefore  $P_s$  random
- Metrics: **outage**, **average**  $P_s$ , combined outage and average.



$$\bar{P}_s = \int P_s(\gamma_s) p(\gamma_s) d\gamma_s$$

# Moment Generating Function Approach

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- Simplifies average  $P_s$  calculation
- Uses alternate Q function representation
- $\overline{P}_s$  reduces to MGF of  $\gamma_s$  distribution
- Closed form or simple numerical calculation for general fading distributions
- Fading greatly increases average  $P_s$ .

# Doppler Effects

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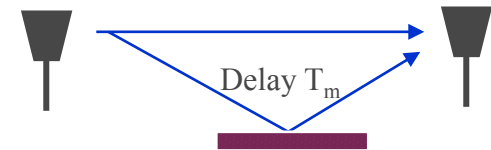
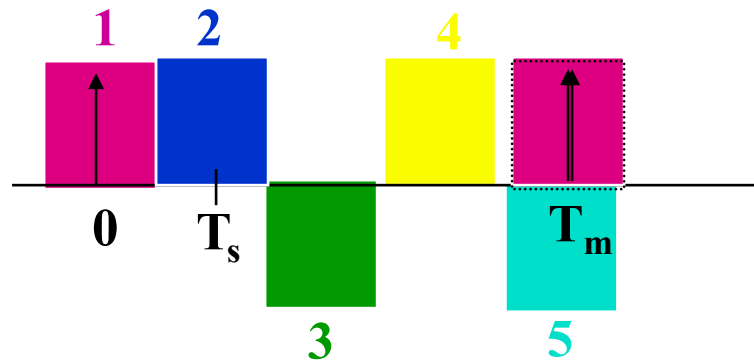
- High doppler causes channel phase to decorrelate between symbols
- Leads to an irreducible error floor for differential modulation
  - Increasing power does not reduce error
- Error floor depends on  $f_D T_b$  as

$$P_{floor} = \frac{1 - J_0(2\pi f_D T_b)}{2} \approx .5(\pi f_D T_b)^2$$



# Delay Spread (ISI) Effects

- Delay spread exceeding a symbol time causes ISI (self interference).



- ISI leads to irreducible error floor:  $\bar{P}_{b, floor} \approx (\sigma_{T_m}/T_s)^2$ 
  - Increasing signal power increases ISI power
- ISI imposes data rate constraint:  $T_s \gg T_m$  ( $R_s \ll B_c$ )

$$R \leq \log_2(M) \times \sqrt{\bar{P}_{b, floor} / \sigma_{T_m}^2}$$

# Diversity

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- Send bits over independent fading paths
  - Combine paths to mitigate fading effects.
- Independent fading paths
  - Space, time, frequency, polarization diversity.
- Combining techniques
  - Selection combining (SC)
  - Maximal ratio combining (MRC)
- Can have diversity at TX or RX
  - In TX diversity, weights constrained by TX power

# Selection Combining

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- Selects the path with the highest gain
- Combiner SNR is the maximum of the branch SNRs.
- CDF easy to obtain ( $\prod_i p(\gamma_i < \gamma_{thr})$ ), pdf found by differentiating the CDF
- $P_{out}$  obtained from CDF. Average  $P_s$  typically found numerically
- Diminishing returns with number of antennas.
- Can get up to about 20 dB of gain.

# MRC and its Performance

- With MRC,  $\gamma_{\Sigma} = \sum \gamma_i$  for branch SNRs  $\gamma_i$ 
  - Optimal technique to maximize output SNR
  - Yields 20-40 dB performance gains
  - Distribution of  $\gamma_{\Sigma}$  hard to obtain

- Standard average BER calculation

$$\bar{P}_s = \int P_s(\gamma_{\Sigma}) p(\gamma_{\Sigma}) d\gamma_{\Sigma} = \int \int \dots \int P_s(\gamma_{\Sigma}) p(\gamma_1) * p(\gamma_2) * \dots * p(\gamma_M) d\gamma_1 d\gamma_2 \dots d\gamma_M$$

- Hard to obtain in closed form
  - Integral often diverges

- MGF Approach: 
$$\bar{P}_s = \frac{\alpha_M}{\pi} \int_0^{\pi/2} \prod_{i=1}^M \mathcal{M}_{\gamma_i} \left[ \frac{-0.5\beta_M}{\sin^2 \phi} \right] d\phi$$

- TX diversity gain with CSI same as RX diversity

# Main Points

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- **Wireless channels introduce path-loss, shadowing and multipath fading**
  - Shadowing introduced outage
  - Flat-fading causes large power fluctuations
  - ISI causes self-interference
- **Performance of digital communications in wireless channels random**
  - Characterized by outage probability and average probability of error in flat-fading
  - Characterized by irreducible error floors in ISI/Doppler
- **Need mechanisms to compensate for multipath**
- **Diversity compensates for effects of flat fading.**