

EE359 – Lecture 9 Outline

- **Announcements:**
 - Project proposals due this Friday at 5pm
 - Midterm will likely be @ Nov. 10, 6-8pm.
 - HWs will be due Friday 5pm SHARP going forward
- Linear Modulation Review
- Linear Modulation Performance in AWGN
- Q-Function representations
- Probability of error in fading
- Outage probability
- Average P_s (P_b)

Passband Modulation Tradeoffs

- Want high rates, high spectral efficiency, high power efficiency, robust to channel, cheap. Our focus
- Amplitude/Phase Modulation (MPSK, MQAM) ↙
 - Information encoded in amplitude/phase
 - More spectrally efficient than frequency modulation
 - Issues: differential encoding, pulse shaping, bit mapping.
- Frequency Modulation (FSK)
 - Information encoded in frequency
 - Continuous phase (CPFSK) special case of FM
 - Bandwidth determined by Carson's rule (pulse shaping)
 - More robust to channel and amplifier nonlinearities

Alternate Q Function Representation

- Traditional Q function representation

$$Q(z) = p(x > z) = \int_z^{\infty} \frac{1}{\sqrt{2\pi}} e^{-x^2/2} dx, \quad x \sim N(0,1)$$
 - Infinite integrand
 - Argument in integral limits
- New representation (Craig'93)

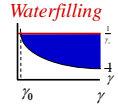
$$Q(z) = \frac{1}{\pi} \int_0^{\pi/2} e^{-z^2 / (\sin^2 \phi)} d\phi$$
 - Leads to closed form solution for P_s in PSK
 - Very useful in fading and diversity analysis

Review of Last Lecture

- Capacity in Flat-Fading: γ known at TX/RX
 - Optimal Rate and Power Adaptation

$$C = \max_{S(\gamma): E[S(\gamma)] = \bar{S}} \int_0^{\infty} B \log_2 \left(1 + \frac{\gamma S(\gamma)}{S} \right) p(\gamma) d\gamma$$

$$\frac{C}{B} = \int_{\gamma_0}^{\infty} \log_2 \left(\frac{\gamma}{\gamma_0} \right) p(\gamma) d\gamma.$$



- Channel Inversion and Truncated Inversion
 - Received SNR constant; Capacity is $B \log_2(1 + \sigma)$ above an outage level associated with truncation
- Capacity of ISI channels
 - Water-filling of power over freq; or time and freq.

Amplitude/Phase Modulation

- Signal over i th symbol period:

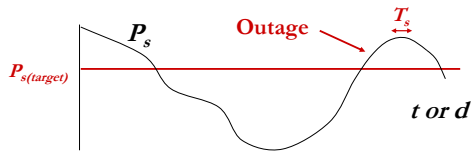
$$s(t) = s_{i1} g(t) \cos(2\pi f_c t + \phi_0) - s_{i2} g(t) \sin(2\pi f_c t + \phi_0)$$

- Pulse shape $g(t)$ typically Nyquist
- Signal constellation defined by (s_{i1}, s_{i2}) pairs
- Can be differentially encoded
- M values for $(s_{i1}, s_{i2}) \Rightarrow \log_2 M$ bits per symbol
- P_s depends on
 - Minimum distance d_{min} (depends on γ_s)
 - # of nearest neighbors α_M
 - Approximate expression: $P_s \approx \alpha_M Q(\sqrt{\beta_M \gamma_s})$

Linear Modulation in Fading

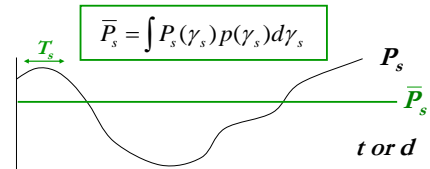
- In fading γ_s and therefore P_s random
- Performance metrics:
 - Outage probability: $p(P_s > P_{target}) = p(\gamma < \gamma_{target})$
 - Average P_s, \bar{P}_s
$$\bar{P}_s = \int_0^{\infty} P_s(\gamma) p(\gamma) d\gamma$$
 - Combined outage and average P_s (next lecture)

Outage Probability



- Probability that P_s is above target
- Equivalently, probability γ_s below target
- Used when $T_c \gg T_s$

Average P_s



- Expected value of random variable P_s
- Used when $T_c \sim T_s$
- Error probability much higher than in AWGN alone
- Alternate Q function approach: $Q(z) = \frac{1}{\pi} \int_0^{\pi/2} e^{-z^2 / (\sin^2 \phi)} d\phi$
 - Simplifies calculations (Get a Laplace Xfm)

Main Points

- Linear modulation more spectrally efficient but less robust than nonlinear modulation
- P_s approximation in AWGN: $P_s \approx \alpha_M Q(\sqrt{\beta_M \gamma_s})$
 - Alternate Q function representation simplifies calculations
- In fading P_s is a random variable, characterized by average value, outage, or combined outage/average
- Outage probability based on target SNR in AWGN.
- Fading greatly increases average P_s .